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## DIRECTPATH™ STEREO LINE DRIVER, ADJUSTABLE GAIN

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### FEATURES

- External Gain Setting Resistors
- Space Saving Package
  - 20-Pin, 4 mm × 4 mm Thin QFN, Thermally Optimized PowerPAD™ Package
- Ground-Referenced Outputs Eliminate DC-Blocking Capacitor
  - Reduce Board Area
  - Reduce Component Cost
  - Improve THD+N Performance
  - No Degradation of Low-Frequency Response Due to Output Capacitors
- Wide Power Supply Range: 1.8 V to 4.5 V
- 2 Vrms/Ch Output Voltage into 600 Ω at 3.3 V supply

- Independent Right and Left Channel Shutdown Control
- Short-Circuit and Thermal Protection
- Pop Reduction Circuitry

### APPLICATIONS

- Set-Top Boxes
- CD / DVD Players
- DVD-Receivers
- HTIB
- PDP / LCD TV's

### DESCRIPTION

The DRV601 is a stereo line driver designed to allow the removal of the output dc-blocking capacitors for reduced component count and cost. The device is ideal for single supply electronics where size and cost are critical design parameters.

The DRV601 is capable of driving 2 Vrms into a 600-Ω load at 3.3 V. The device has external gain setting resistors, that support a gain range of -1V/V to -10V/V, and line outputs that has ±8-kV IEC ESD protection. The device has independent shutdown control for the right and left audio channels.

The DRV601 is available in a 4 mm × 4 mm Thin QFN package.



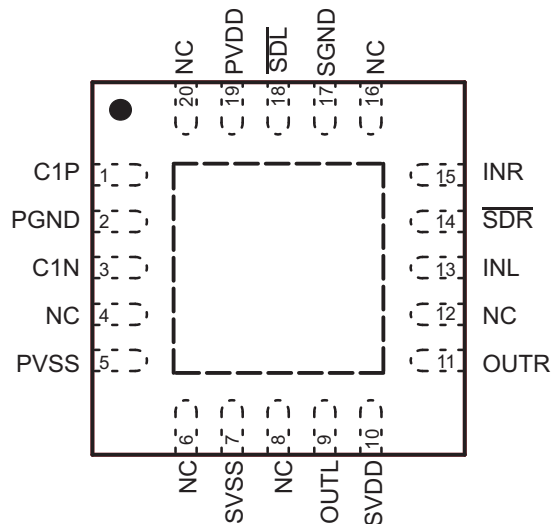
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These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

### RTJ (QFN) PACKAGE (TOP VIEW)



NC – No internal connection

### TERMINAL FUNCTIONS

TERMINAL		I/O	DESCRIPTION
NAME	QFN		
C1P	1	I/O	Charge pump flying capacitor positive terminal
PGND	2	I	Power ground, connect to ground.
C1N	3	I/O	Charge pump flying capacitor negative terminal
NC	4, 6, 8, 12, 16, 20		No connection
PVSS	5	O	Output from charge pump.
SVSS	7	I	Amplifier negative supply, connect to PVSS via star connection.
OUTL	9	O	Left audio channel output signal
SVDD	10	I	Amplifier positive supply, connect to PVDD via star connection.
OUTR	11	O	Right audio channel output signal
INL	13	I	Left audio channel input signal
$\overline{\text{SDR}}$	14	I	Right channel shutdown, active low logic.
INR	15	I	Right audio channel input signal
SGND	17	I	Signal ground, connect to ground.
$\overline{\text{SDL}}$	18	I	Left channel shutdown, active low logic.
PVDD	19	I	Supply voltage, connect to positive supply.
Exposed Pad			Exposed pad must be soldered to a floating plane. Do NOT connect to power or ground.

## ABSOLUTE MAXIMUM RATINGS<sup>(1)</sup>

over operating free-air temperature range,  $T_A = 25^\circ\text{C}$  (unless otherwise noted)

	VALUE / UNIT
Supply voltage, AVDD, PVDD	–0.3 V to 5.5 V
$V_I$ Input voltage	$V_{SS} - 0.3\text{ V}$ to $V_{DD} + 0.3\text{ V}$
$R_{(Load)}$ Minimum load impedance	$\geq 100\ \Omega$
$T_A$ Operating free-air temperature range	$0^\circ\text{C}$ to $70^\circ\text{C}$
$T_J$ Operating junction temperature range	$0^\circ\text{C}$ to $150^\circ\text{C}$
$T_{stg}$ Storage temperature range	$-65^\circ\text{C}$ to $85^\circ\text{C}$

- (1) Stresses beyond those listed under *absolute maximum ratings* may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under *recommended operating conditions* is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

## AVAILABLE OPTIONS

$T_A$	PACKAGED DEVICES <sup>(1)</sup>	PART NUMBER	SYMBOL
$-40^\circ\text{C}$ to $85^\circ\text{C}$	20-pin, 4 mm × 4 mm QFN	DRV601RTJ <sup>(2)</sup>	AKQ

- (1) For the most current package and ordering information, see the Package Option Addendum at the end of this document, or see the TI website at [www.ti.com](http://www.ti.com).  
(2) The RTJ package is only available taped and reeled. To order, add the suffix "R" to the end of the part number for a reel of 3000, or add the suffix "T" to the end of the part number for a reel of 250 (e.g., DRV601RTJR).

## RECOMMENDED OPERATING CONDITIONS

	MIN	MAX	UNIT
$V_{SS}$ Supply voltage, AVDD, PVDD	1.8	4.5 <sup>(1)</sup>	V
$V_{IH}$ High-level input voltage	$\overline{SDL}$ , $\overline{SDR}$	1.5	V
$V_{IL}$ Low-level input voltage	$\overline{SDL}$ , $\overline{SDR}$	0.5	V
$T_A$ Operating free-air temperature	0	70	$^\circ\text{C}$

- (1) Device can shut down for  $V_{DD} > 4.5\text{ V}$  to prevent damage to the device.

## ELECTRICAL CHARACTERISTICS

$T_A = 25^\circ\text{C}$  (unless otherwise noted)

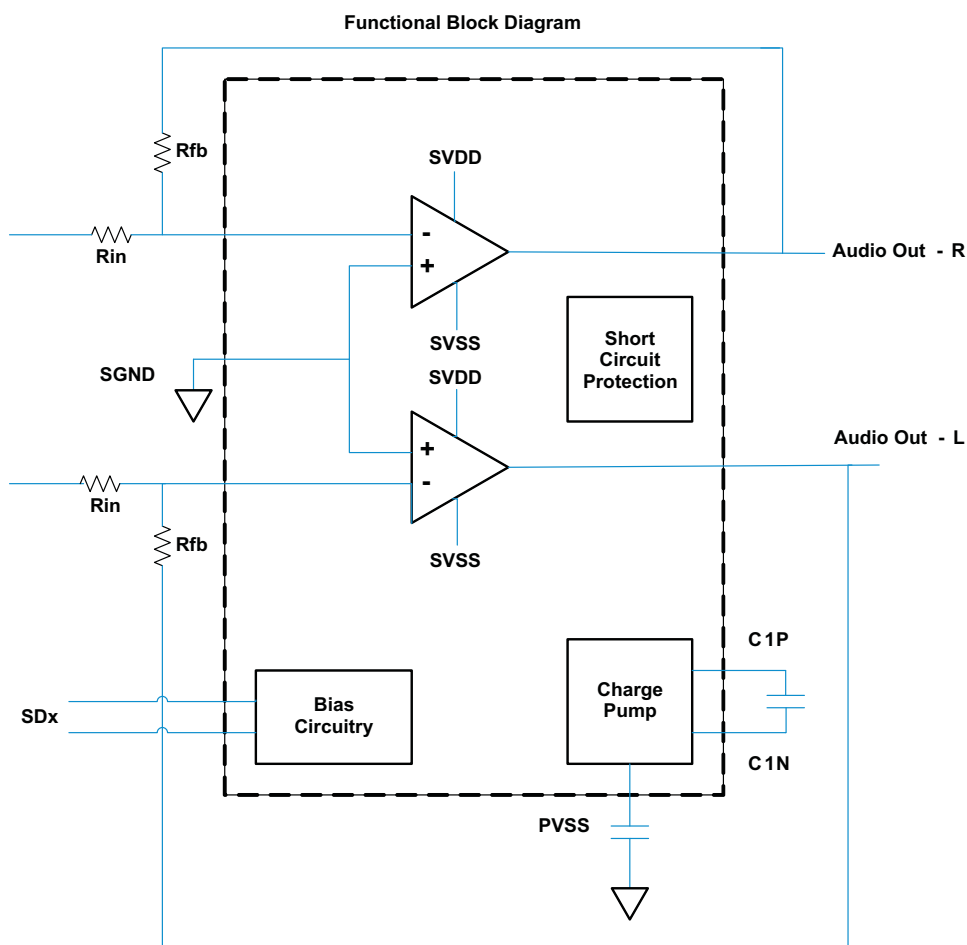
PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
$ V_{OS} $ Output offset voltage	$V_{DD} = 1.8\text{ V}$ to $4.5\text{ V}$ , Inputs grounded			8	mV
PSRR Power Supply Rejection Ratio	$V_{DD} = 1.8\text{ V}$ to $4.5\text{ V}$		88		dB
$V_{OH}$ High-level output voltage	$V_{DD} = 3.3\text{ V}$ , $R_L = 600\ \Omega$	3.10			V
$V_{OL}$ Low-level output voltage	$V_{DD} = 3.3\text{ V}$ , $R_L = 600\ \Omega$			–3.05	V
$ I_{IH} $ High-level input current ( $\overline{SDL}$ , $\overline{SDR}$ )	$V_{DD} = 4.5\text{ V}$ , $V_I = V_{DD}$			1	$\mu\text{A}$
$ I_{IL} $ Low-level input current ( $\overline{SDL}$ , $\overline{SDR}$ )	$V_{DD} = 4.5\text{ V}$ , $V_I = 0\text{ V}$			1	$\mu\text{A}$
$I_{DD}$ Supply Current	$V_{DD} = 1.8\text{ V}$ , No load, $\overline{SDL} = \overline{SDR} = V_{DD}$		5.3		mA
	$V_{DD} = 3.3\text{ V}$ , No load, $\overline{SDL} = \overline{SDR} = V_{DD}$		7.1		
	$V_{DD} = 4.5\text{ V}$ , No load, $\overline{SDL} = \overline{SDR} = V_{DD}$		8.7		
	Shutdown mode, $V_{DD} = 1.8\text{ V}$ to $4.5\text{ V}$			1	$\mu\text{A}$

## OPERATING CHARACTERISTICS

$V_{DD} = 3.3\text{ V}$ ,  $T_A = 25^\circ\text{C}$ ,  $R_L = 600\ \Omega$ ,  $C_{(PUMP)} = C_{(PVSS)} = 1\ \mu\text{F}$ ,  $C_{IN} = 1\ \mu\text{F}$ ,  $R_{in} = 10\ \text{k}\Omega$ ,  $R_{fb} = 20\ \text{k}\Omega$  (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
$V_O$	Output Voltage(Outputs In Phase)	THD = 1%, $V_{DD} = 3.3\text{ V}$ , $f = 1\text{ kHz}$		2.1		$V_{RMS}$
		THD = 1%, $V_{DD} = 4.5\text{ V}$ , $f = 1\text{ kHz}$		2.7		
		THD = 1%, $V_{DD} = 4.5\text{ V}$ , $f = 1\text{ kHz}$ , $R_L = 100\ \text{k}\Omega$		2.8		
THD+N	Total harmonic distortion plus noise	$V_O = 2\text{ Vrms}$ , $f = 1\text{ kHz}$		0.008%		
		$V_O = 2\text{ Vrms}$ , $f = 6.67\text{ kHz}$		0.030%		
	Crosstalk	$V_O = 2\text{ Vrms}$ , $f = 1\text{ kHz}$		-80		dB
$A_{vo}$	Open-loop voltage gain			155		dB
$R_{in}$	Input resistor range		1	10	47	k $\Omega$
$R_{fb}$	Feedback resistor range		4.7	20	100	k $\Omega$
	Slew rate			2.2		V/ $\mu\text{s}$
	Maximum capacitive load			300		pF
$V_n$	Noise output voltage	22-kHz filter, A-weighted		10		$\mu\text{Vrms}$
ESD	Electrostatic discharge	OUTR, OUTL		$\pm 8$		kV
$f_{osc}$	Charge pump switching frequency		280	320	420	kHz
	Start-up time from shutdown			450		$\mu\text{s}$
	Input impedance		1			M $\Omega$
SNR	Signal-to-noise ratio	$V_O = 2\text{ Vrms}$ (THD+N = 0.1%), 22-kHz BW, A-weighted		105		dB
$G_{(bw)}$	Unity Gain Bandwidth			3.5		MHz
	Thermal shutdown	Threshold	150		170	$^\circ\text{C}$
		Hysteresis		15		$^\circ\text{C}$

## Functional Block Diagram



## TYPICAL CHARACTERISTICS

 $C_{(PUMP)} = C_{(PVSS)} = 1\ \mu\text{F}$ ,  $C_{IN} = 1\ \mu\text{F}$ ,  $R_{in} = 10\ \text{k}\Omega$ ,  $R_{fb} = 20\ \text{k}\Omega$  (unless otherwise noted)

## Table of Graphs

		FIGURE
Total harmonic distortion + noise	vs Output Voltage	1-6
Total harmonic distortion + noise	vs Frequency	7-8
Quiescent supply current	vs Supply voltage	9
Output spectrum		10
Gain and phase	vs Frequency	11-12

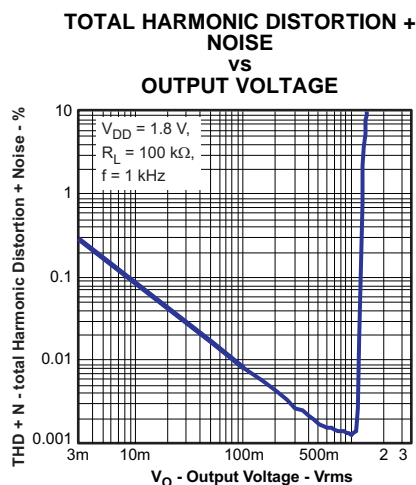


Figure 1.

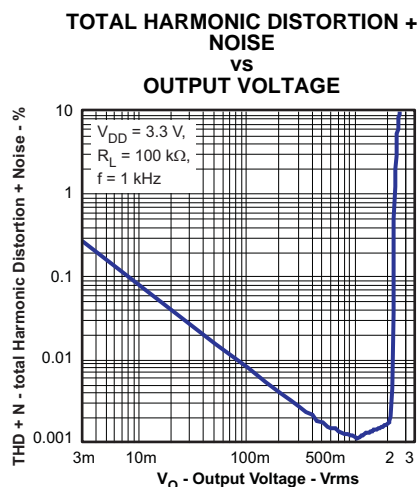


Figure 2.

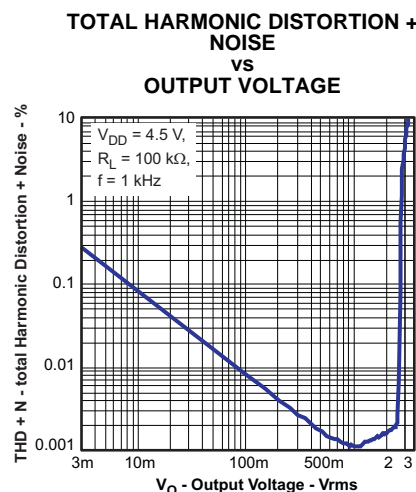


Figure 3.

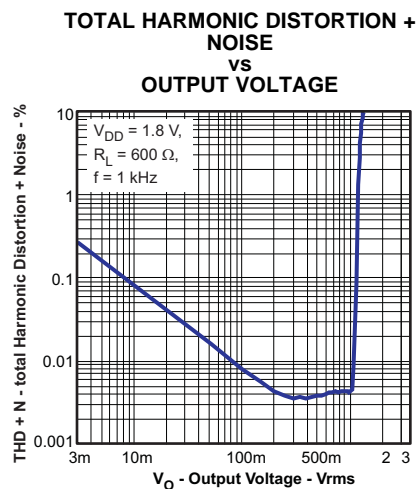


Figure 4.

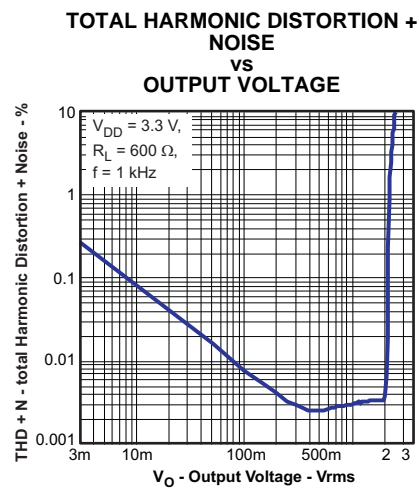


Figure 5.

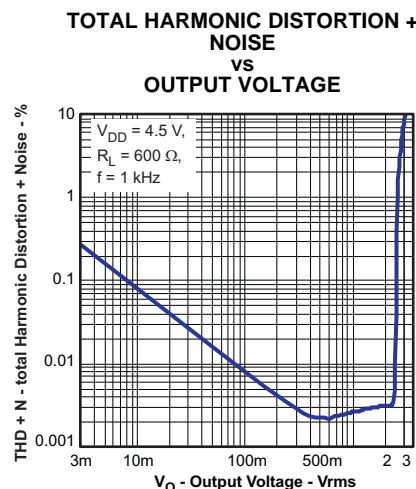


Figure 6.

**TOTAL HARMONIC DISTORTION +  
NOISE  
VS  
FREQUENCY**

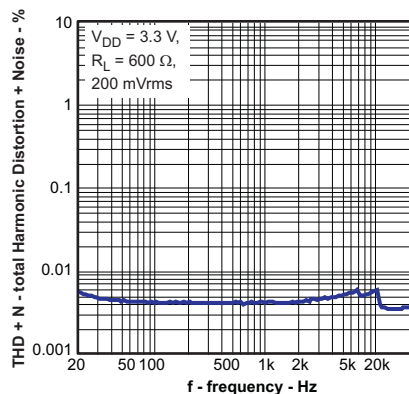


Figure 7.

**TOTAL HARMONIC DISTORTION +  
NOISE  
VS  
FREQUENCY**

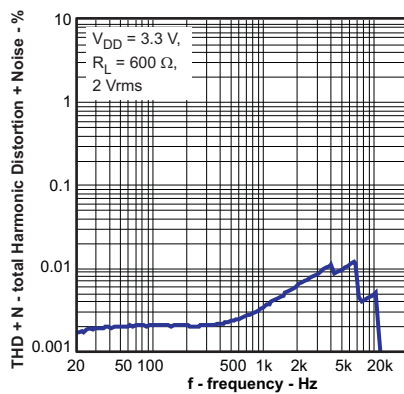


Figure 8.

**QUIESCENT SUPPLY CURRENT  
VS  
SUPPLY VOLTAGE**

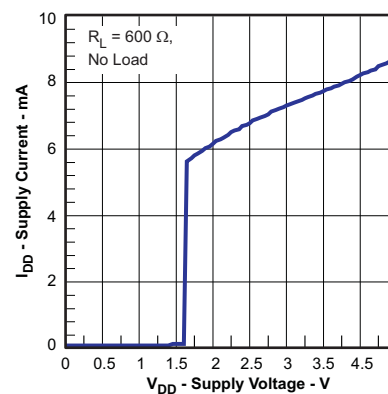


Figure 9.

**FFT  
VS  
FREQUENCY**

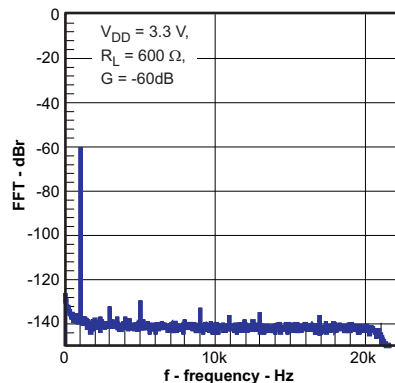


Figure 10.

**GAIN  
VS  
FREQUENCY**

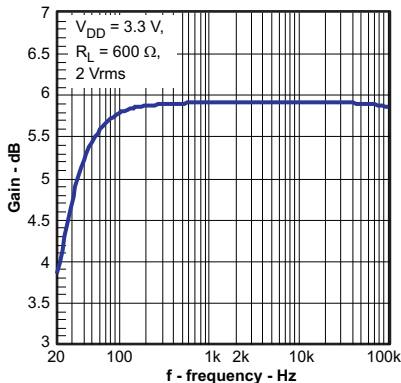


Figure 11.

**PHASE  
VS  
FREQUENCY**

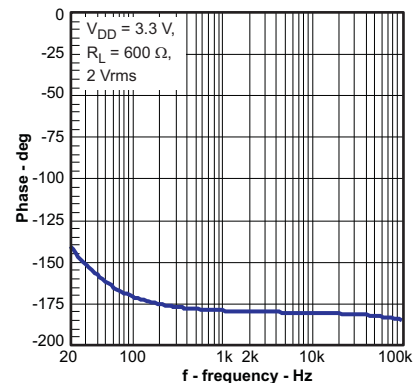


Figure 12.

## APPLICATION INFORMATION

### Line Driver Amplifiers

Single-supply Line Driver amplifiers typically require dc-blocking capacitors. The top drawing in [Figure 13](#) illustrates the conventional Line Driver amplifier connection to the load and output signal.

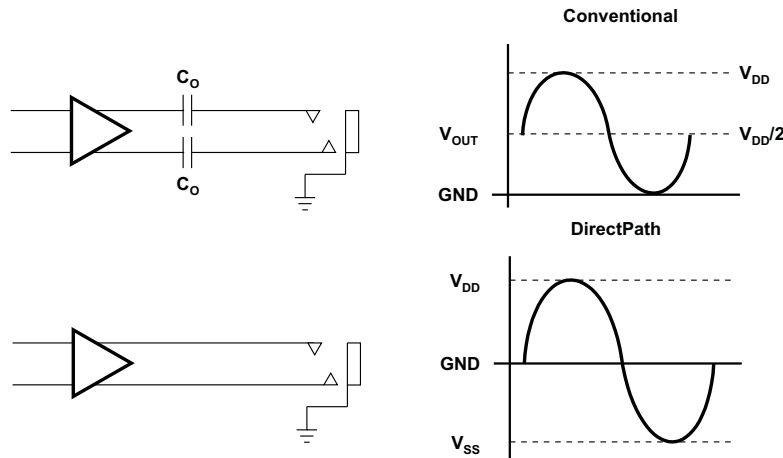
DC blocking capacitors are often large in value. The line load (typical resistive values of 600  $\Omega$  to 10 k $\Omega$ ) combine with the dc blocking capacitors to form a high-pass filter. [Equation 1](#) shows the relationship between the load impedance ( $R_L$ ), the capacitor ( $C_O$ ), and the cutoff frequency ( $f_c$ ).

$$f_c = \frac{1}{2\pi R_L C_O} \quad (1)$$

$C_O$  can be determined using [Equation 2](#), where the load impedance and the cutoff frequency are known.

$$C_O = \frac{1}{2\pi R_L f_c} \quad (2)$$

If  $f_c$  is low, the capacitor must then have a large value because the load resistance is small. Large capacitance values require large package sizes. Large package sizes consume PCB area, stand high above the PCB, increase cost of assembly, and can reduce the fidelity of the audio output signal.



**Figure 13. Amplifier Applications**

The DirectPath™ amplifier architecture operates from a single supply but makes use of an internal charge pump to provide a negative voltage rail. Combining the user provided positive rail and the negative rail generated by the IC, the device operates in what is effectively a split supply mode. The output voltages are now centered at zero volts with the capability to swing to the positive rail or negative rail. The DirectPath™ amplifier requires no output dc blocking capacitors. The bottom block diagram and waveform of [Figure 13](#) illustrate the ground-referenced Line Driver architecture. This is the architecture of the DRV601.



## Charge Pump Flying Capacitor and PVSS Capacitor

The charge pump flying capacitor serves to transfer charge during the generation of the negative supply voltage. The PVSS capacitor must be at least equal to the charge pump capacitor in order to allow maximum charge transfer. Low ESR capacitors are an ideal selection, and a value of 1µF is typical. Capacitor values that are smaller than 1µF can be used, but the maximum output voltage may be reduced and the device may not operate to specifications.

## Decoupling Capacitors

The DRV601 is a DirectPath™ Line Driver amplifier that require adequate power supply decoupling to ensure that the noise and total harmonic distortion (THD) are low. A good low equivalent-series-resistance (ESR) ceramic capacitor, typically 2.2µF, placed as close as possible to the device V<sub>DD</sub> lead works best. Placing this decoupling capacitor close to the DRV601 is important for the performance of the amplifier. For filtering lower frequency noise signals, a 10-µF or greater capacitor placed near the audio power amplifier would also help, but it is not required in most applications because of the high PSRR of this device.

## Gain setting resistors ranges

The gain setting resistors, R<sub>in</sub> and R<sub>fb</sub>, must be chosen so that noise, stability and input capacitor size of the DRV601 is kept within acceptable limits. Voltage gain is defined as R<sub>fb</sub> divided by R<sub>in</sub>.

Selecting values that are too low demands a large input ac-coupling capacitor, C<sub>IN</sub>. Selecting values that are too high increases the noise of the amplifier. [Table 1](#) lists the recommended resistor values for different gain settings.

**Table 1. Recommended Resistor Values**

Gain	Input Resistor Value, R <sub>in</sub>	Feedback Resistor Value, R <sub>fb</sub>
-1 V/V	10 kΩ	10 kΩ
-1.5 V/V	10 kΩ	15 kΩ
-2 V/V	10 kΩ	20 kΩ
-10 V/V	4,7 kΩ	47 kΩ

## Input-Blocking Capacitors

DC input-blocking capacitors are required to be added in series with the audio signal into the input pins of the DRV601. These capacitors block the DC portion of the audio source and allow the DRV601 inputs to be properly biased to provide maximum performance.

These capacitors form a high-pass filter with the input resistor, R<sub>in</sub>. The cutoff frequency is calculated using [Equation 3](#). For this calculation, the capacitance used is the input-blocking capacitor and the resistance is the input resistor chosen from the gain table above, then the frequency and/or capacitance can be determined when one of the two values are given.

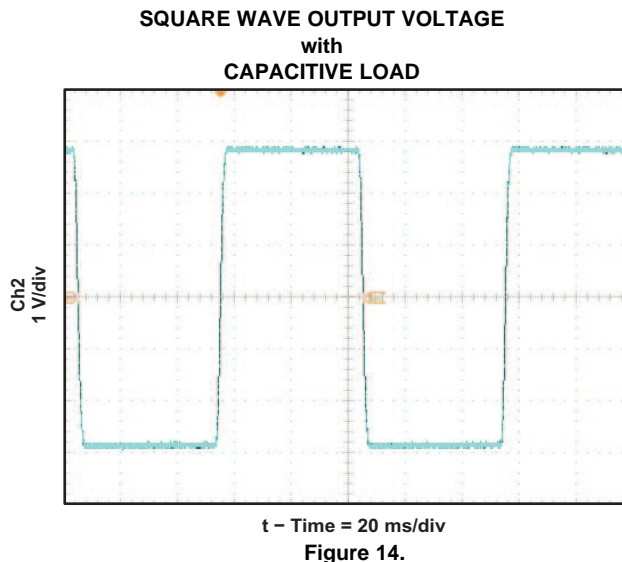
$$f_{C_{IN}} = \frac{1}{2\pi R_{IN} C_{IN}} \quad \text{or} \quad C_{IN} = \frac{1}{2\pi f_{C_{IN}} R_{IN}} \quad (3)$$

## Supply Voltage Limiting At 4.5 V

The DRV601 have a built-in charge pump which serves to generate a negative rail for the line driver. Because the line driver operates from a positive voltage and negative voltage supply, circuitry has been implemented to protect the devices in the amplifier from an overvoltage condition. Once the supply is above 4.5 V, the DRV601 can shut down in an overvoltage protection mode to prevent damage to the device. The DRV601 resume normal operation once the supply is reduced to 4.5 V or lower.

## Capacitive load

The DRV601 has the ability to drive a high capacitive load up to 330pF directly, higher capacitive loads can be accepted by adding a series resistor of 10Ω or larger. The figure below shows a 10kHz signal into a 470pF capacitor using the 10R series resistor.



## Layout Recommendations

A proposed layout for the DRV601 can be seen in the DRV601EVM user's guide, [SLOU215](#), and the Gerber files can be downloaded on [www.ti.com](http://www.ti.com), open the DRV601 product folder and look in the Tools & Software folder.

## Exposed Pad On DRV601RTJ Package

The exposed metal pad on the DRV601RTJ package must be soldered down to a pad on the PCB in order to maintain reliability. *The pad on the PCB should be allowed to float and not be connected to ground or power.* Connecting this pad to power or ground prevents the device from working properly because it is connected internally to PVSS.

## SGND and PGND Connections

The SGND and PGND pins of the DRV601 must be routed back to the decoupling capacitor separately in order to provide proper device operation. If the SGND and PGND pins are connected directly to each other, the part functions without risk of failure, but the noise and THD performance do not meet the specifications.

## Gain setting resistors

The gain setting resistors,  $R_{in}$  and  $R_{fb}$ , must be placed close to pin 13 respectively pin 17 to minimize the capacitive loading on these input pins and to ensure maximum stability of the DRV601. For the recommended PCB layout, see the DRV601EVM user guide.

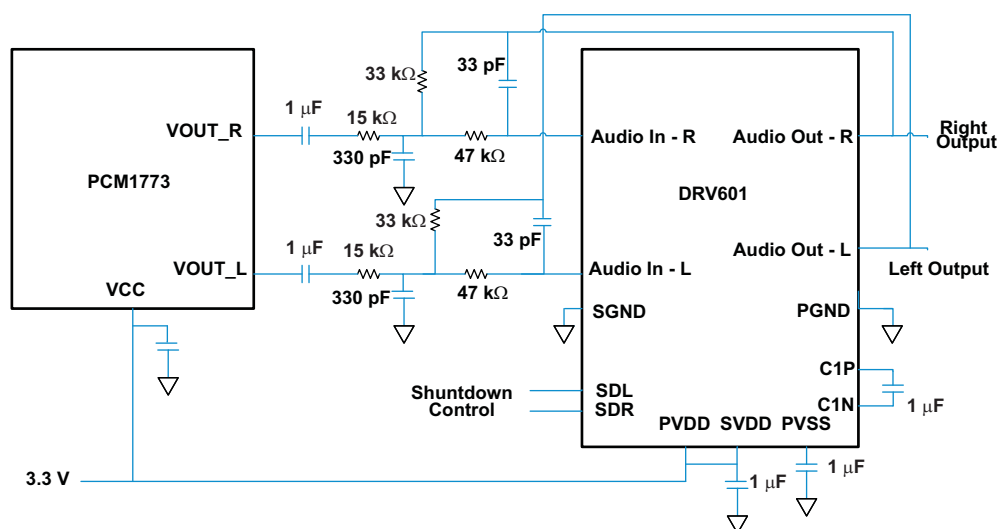


Figure 15. Application Circuit

## PACKAGING INFORMATION

Orderable Device	Status <sup>(1)</sup>	Package Type	Package Drawing	Pins	Package Qty	Eco Plan <sup>(2)</sup>	Lead/Ball Finish	MSL Peak Temp <sup>(3)</sup>
DRV601RTJR	ACTIVE	QFN	RTJ	20	3000	Green (RoHS & no Sb/Br)	Cu NiPdAu	Level-2-260C-1 YEAR
DRV601RTJRG4	ACTIVE	QFN	RTJ	20	3000	Green (RoHS & no Sb/Br)	Cu NiPdAu	Level-2-260C-1 YEAR
DRV601RTJT	ACTIVE	QFN	RTJ	20	250	Green (RoHS & no Sb/Br)	Cu NiPdAu	Level-2-260C-1 YEAR
DRV601RTJTG4	ACTIVE	QFN	RTJ	20	250	Green (RoHS & no Sb/Br)	Cu NiPdAu	Level-2-260C-1 YEAR

<sup>(1)</sup> The marketing status values are defined as follows:

**ACTIVE:** Product device recommended for new designs.

**LIFEBUY:** TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

**NRND:** Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

**PREVIEW:** Device has been announced but is not in production. Samples may or may not be available.

**OBSOLETE:** TI has discontinued the production of the device.

<sup>(2)</sup> Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check <http://www.ti.com/productcontent> for the latest availability information and additional product content details.

**TBD:** The Pb-Free/Green conversion plan has not been defined.

**Pb-Free (RoHS):** TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.

**Pb-Free (RoHS Exempt):** This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

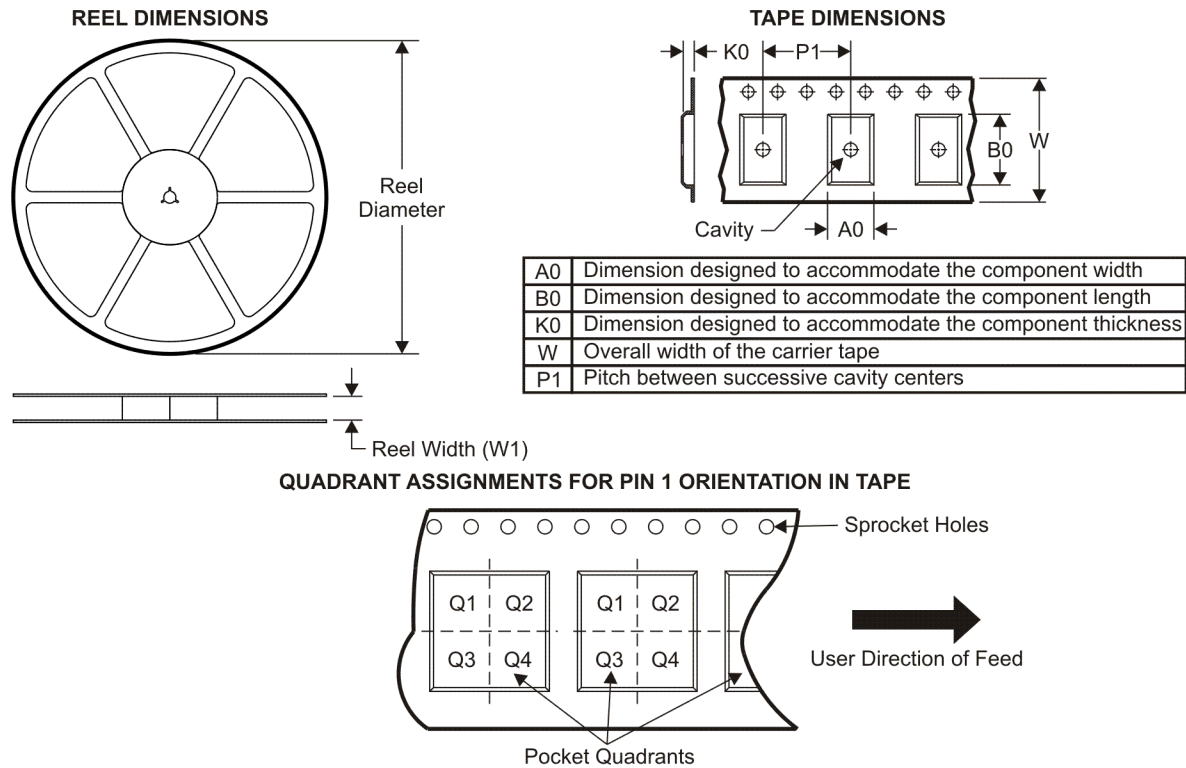
**Green (RoHS & no Sb/Br):** TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

<sup>(3)</sup> MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

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**TAPE AND REEL INFORMATION**



\*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
DRV601RTJT	QFN	RTJ	20	250	180.0	12.4	4.3	4.3	1.5	8.0	12.0	Q2

## TAPE AND REEL BOX DIMENSIONS

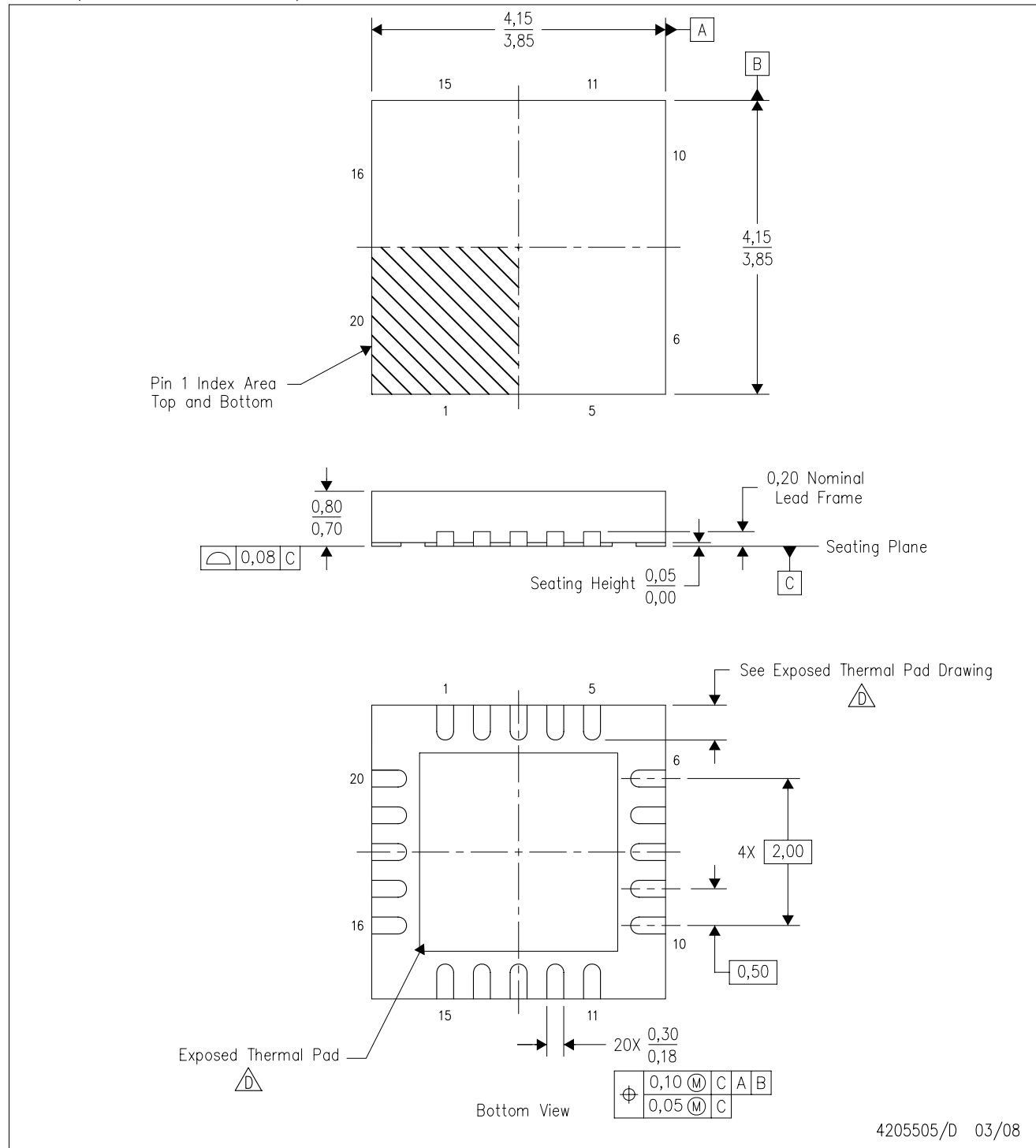



\*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
DRV601RTJT	QFN	RTJ	20	250	190.5	212.7	31.8

RTJ (S-PWQFN-N20)

PLASTIC QUAD FLATPACK NO-LEAD



- NOTES:
- A. All linear dimensions are in millimeters. Dimensioning and tolerancing per ASME Y14.5-1994.
  - B. This drawing is subject to change without notice.
  - C. QFN (Quad Flatpack No-Lead) package configuration.
  -  The package thermal pad must be soldered to the board for thermal and mechanical performance. See the Product Data Sheet for details regarding the exposed thermal pad dimensions.

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